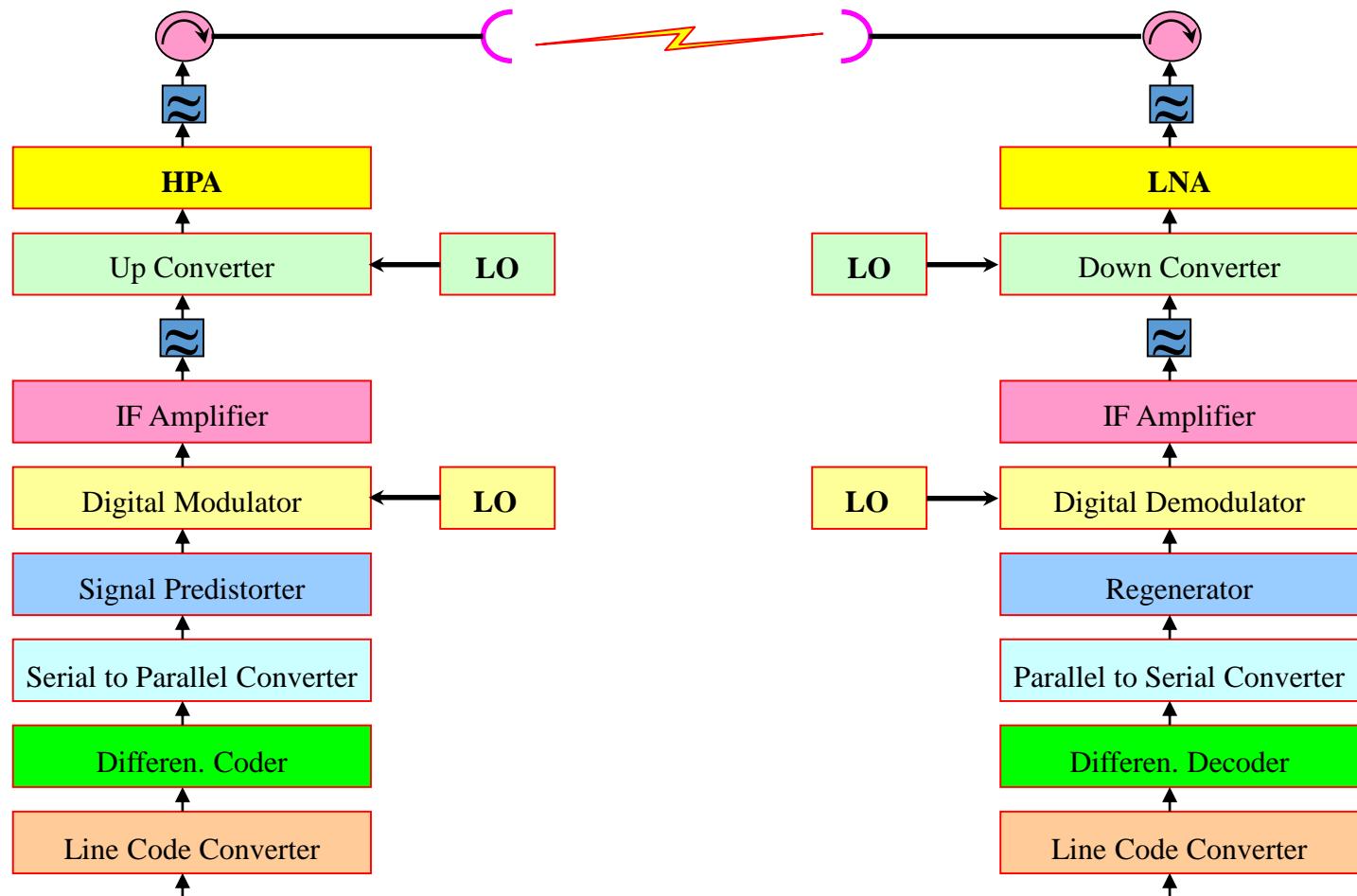


UNJUK KERJA

REF : FREEMAN

BLOK SISTEM KOMUNIKASI RADIO DIGITAL



Energy per bit per Noise density ratio Eb/No

Efisiensi pd siskom digital yg bernoise dinyatakan dengan Eb/No yg diterima utk mendapatkan laju error yg ditentukan.

Eb : Energi per bit

$$E_{b(watt)} = \frac{RSL_{(watt)}}{R_{(bit/det)}}$$

$$E_{b(dBw)} = RSL_{(dBw)} - 10\log R_{(bit/det)}$$

No : aras noise pd penerima sempurna (dBw atau dBm) + noise figure perangkat (dB)

$$N_{o(dBw)} = -204_{(dBw)} + NF_{(dB)}$$

$$\frac{E_b}{N_o} = RSL_{(dBw)} - 10\log B_{(bit/det)} + 204_{(dBw)} - NF_{(dB)}$$

Isu Regulasi - FCC

- Laju bit (bps) harus \geq lebar pita (Hz) yg ditetapkan, kecuali lebar pita yg digunakan menghitung laju minimal tidak termasuk pita pengaman.
- Perangkat yg digunakan utk trans suara yg sudah beroperasi atau yg diaplikasikan sebelum 1 juni 1997:
 - di pita 2110 – 2130 MHz dan 2160 – 2180 MHz lebar pita yg dialokasikan harus mampu menyalurkan minimal 96 kanal.
 - di pita 3700 – 4200 MHz , 5925 – 6425 MHz (lebar pita 30 MHz), 10700 – 11700 MHz (lebar pita 30 dan 40 MHz) harus mampu menyalurkan 1152 kanal
- Kapasitas dan persyaratan pembebanan harus memenuhi perangkat yg digunakan dan beroperasi setelah 1 juni 1997 pada pita 3700 – 4200 MHz , 5925 – 6425 MHz dan 6525 – 6875 MHz (6GHz), 10550 – 10680 MHz (10 GHz) dan 10700 – 11700 MHz (11 GHz) seperti pd tabel berikut :

| Nominal channel bandwidth (MHz) | Minimum payload capacity (Mbits/s) | Minimum traffic loading payload (as percent of payload capacity) | Typical utilization |
|---------------------------------|------------------------------------|--|---------------------|
| 0.400 | 1.54 | N/A | 1 DS-1 |
| 0.800 | 3.08 | N/A | 2 DS-1 |
| 1.25 | 3.08 | N/A | 2 DS-1 |
| 1.60 | 6.17 | N/A | 4 DS-1 |
| 2.50 | 6.17 | N/A | 4 DS-1 |
| 3.75 | 12.3 | N/A | 8 DS-1 |
| 5.0 | 18.5 | N/A | 12 DS-1 |
| 10.0 | 44.7 | 50 ³ | 1 DS-3/STS-1 |
| 20.0 | 89.4 | 50 ³ | 2 DS-3/STS-1 |
| 30.0 (11 GHz) | 89.4 | 50 ³ | 2 DS-3/STS-1 |
| 30.0 (5 GHz) | 134.1 | 50 ³ | 3 DS-3/STS-1 |
| 40 | 134.1 | 50 ³ | 3 DS-3/STS-1 |

Pembatasan Emisi - FCC

- Pd frek < 15 GHz, dlm setiap pita 4 KHz, frek tengah yg menyimpang dr frekuensi yg ditetapkan 50 % < Δf =< 250 % berlaku persamaan berikut tapi tidak boleh kurang dari 50 dB.

$$A = 35 + 0,8(P - 50) + 10\log_{10} B$$

- Redaman > 80 dB atau daya absolut < -13 dBm/1MHz tidak diperlukan.
- A : redaman (dB) dibawah aras daya keluaran rata-rata
- P : prosen penyimpangan dr frek tengah lebar pita transmisi
- B : lebar pita yg dialokasikan (MHz)
- Pd frek > 15 GHz, dlm setiap pita 1 MHz, frek tengah yg menyimpang dari frek yg dialokasikan 50 % < Δf =< 250 % dr lebar pita yg ditetapkan berlaku persamaan sbb, tetapi tidak kurang dr 11 dB.

$$A = 11 + 0,4(P - 50) + 10\log_{10} B$$

- Redaman > 56 dB atau daya absolut < -13 dBm tidak diperlukan.

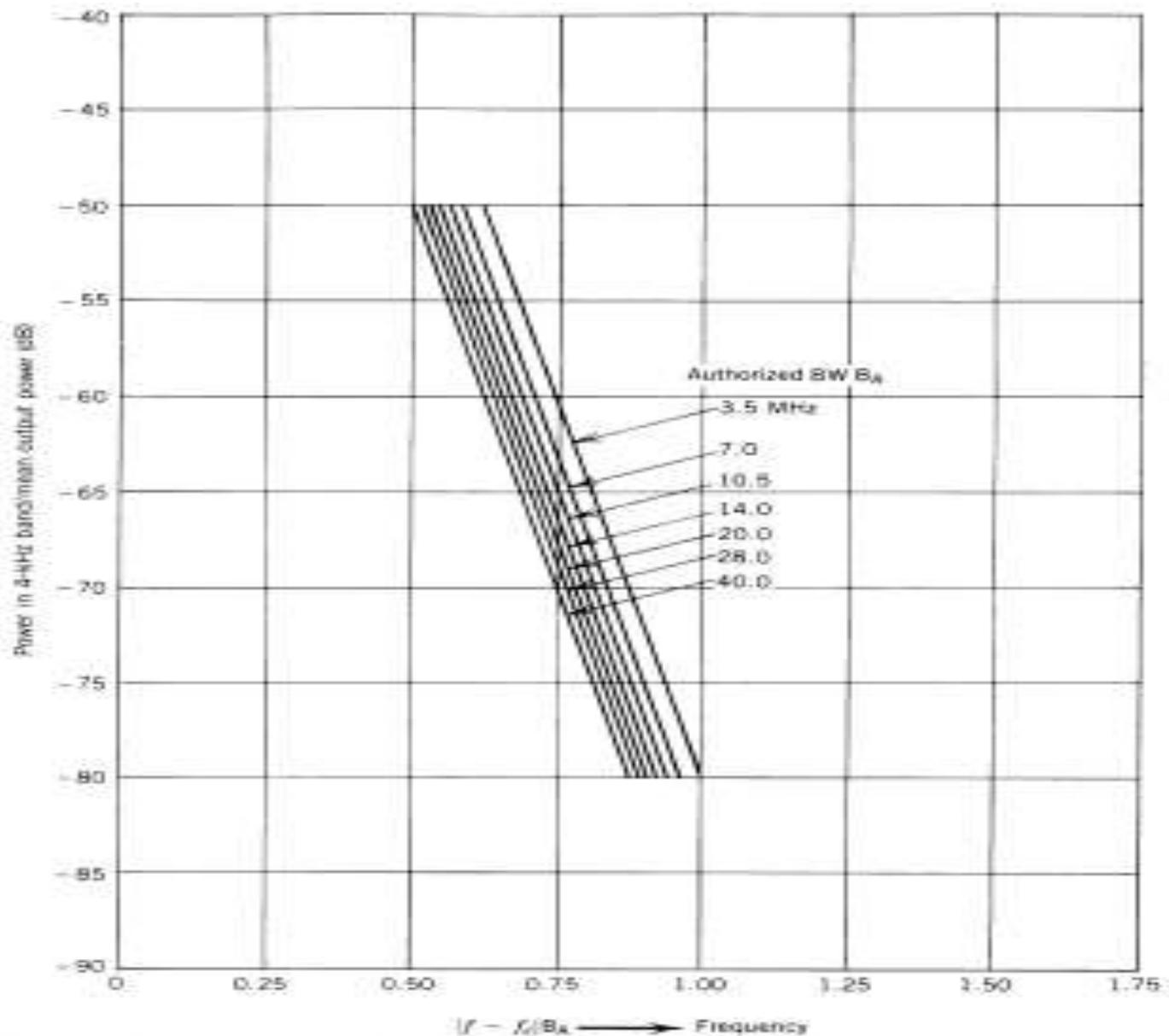


Figure 3.1. FCC Docket 19311 Spectrum Masks: f_c = center frequency; f = frequency of interest. (From Ref. 3.)

Bit Packing

- Kadang2 memperkirakan lebar pita yg diperlukan pd transmisi biner 1 bit/Hz,
- Pd BPSK, fasa pertama menyatakan binary 1 dan fasa kedua menyatakan binary 0, jarak keduanya 180° .
- Pd QPSK, terdapat 4 state fasa, yi $0^\circ = 0,1; 90^\circ = 0,0; 180^\circ = 1,1$ dan $270^\circ = 1,0$.
- Point penting : lebar pita fungsi dari transisi per detik atau perubahan state per detik
- Secara teori QPSK dapat menampung bit packing 2 bit/Hz.
- 8-PSK dapat menampung 3 bit/Hz
- 16-PSK dapat menampung 4 bit/Hz

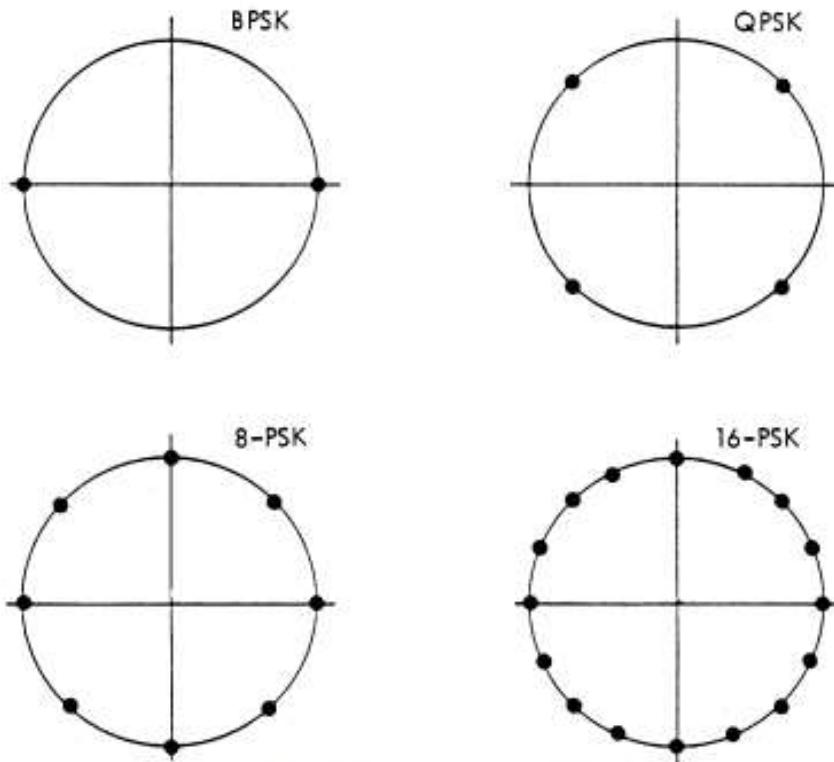


Figure 3.3. Signal state diagrams for PSK.

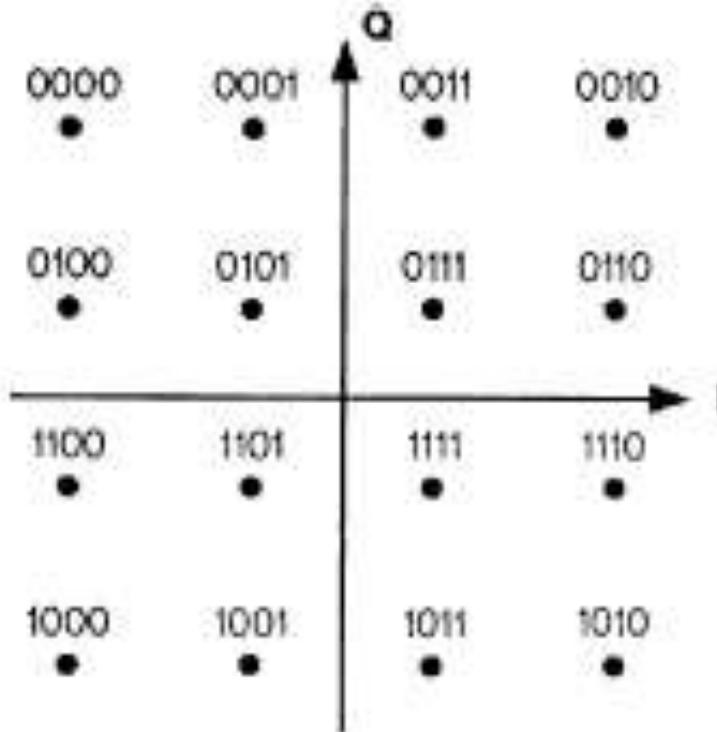


Figure 3.4. 16-QAM state diagram. I = in-phase, Q = quadrature.

Aras amplitudo +3, +1, -1, -3

Bit packing 16 QAM = 16-ary PSK = 4 bit/Hz

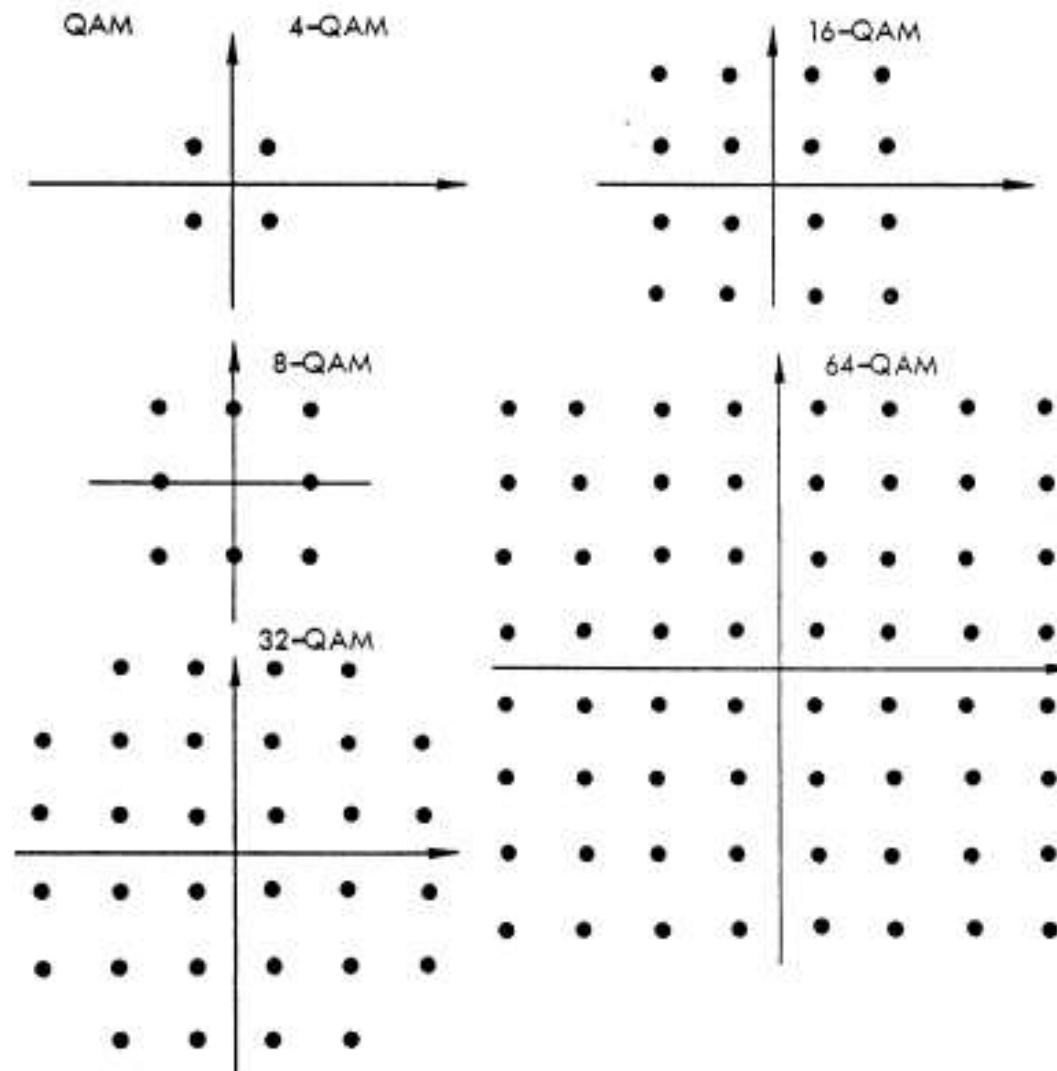


Figure 3.5. Signal state diagrams for 4-, 8-, 16-, 32-, and 64-QAM.

Efisiensi Spektral

Hubungan laju bit R_b dengan laju simbol (1/T) dan jumlah titik level modulasi M

$$R_{b(bps)} = \left(\frac{1}{T} \right)_{ps} \log_2 M$$

Efisiensi spektral η

$$\eta_{(bit/Hz)} = \frac{R_b}{W} = \frac{1}{WT} \log_2 M$$

W : lebar pita RF

Tanpa interferensi kanal → WT dapat = 1

Lebar pita Nyquist → W = 1/T

Lebar pita Nyquist + lebar pita ekses W = (1+α)/T

α : cosine roll off factor, umumnya 0,5

Utk $1/T = \frac{3}{4} W \rightarrow \eta_{(bit/Hz)} = \frac{3}{4} \log_2 M$

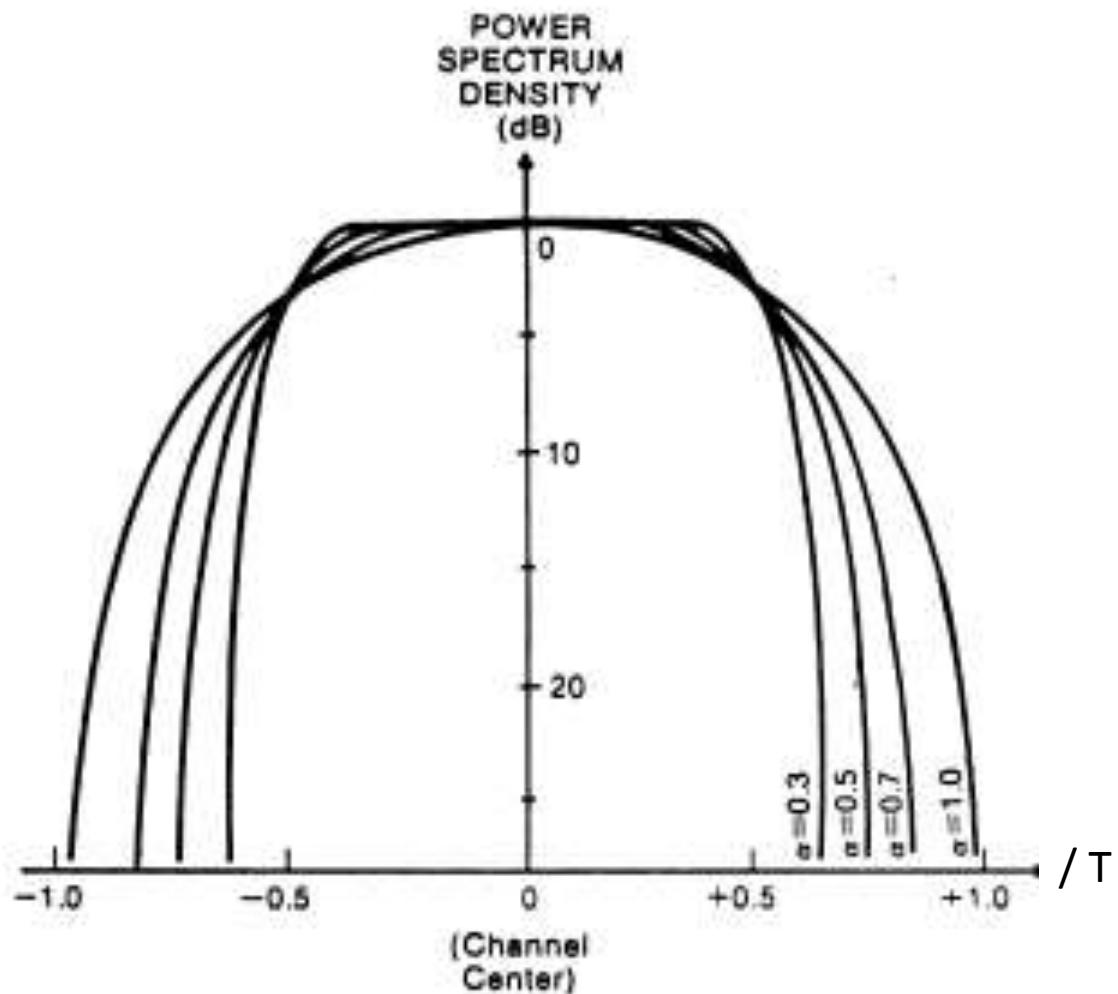


Figure 3.6. Transmitted spectra for typical cosine rolloff shaping. (From *IEEE Communications Magazine*, Oct. 1986, Figure 5, reprinted with permission; Ref. 4.)

Distorsi Penguat Daya

- Penguat daya pemancar gel mikro mrpk perangkat peak-power limited → perangkat menjadi tidak linier jika mendekati saturasi → perlu diperhatikan distorsi secara serius utk gel M-QAM.
- Jika daya puncak berada pd daerah non linier → distorsi non linier
- Pd penerima diperlukan daya rata2 utk mendapatkan BER tertentu.

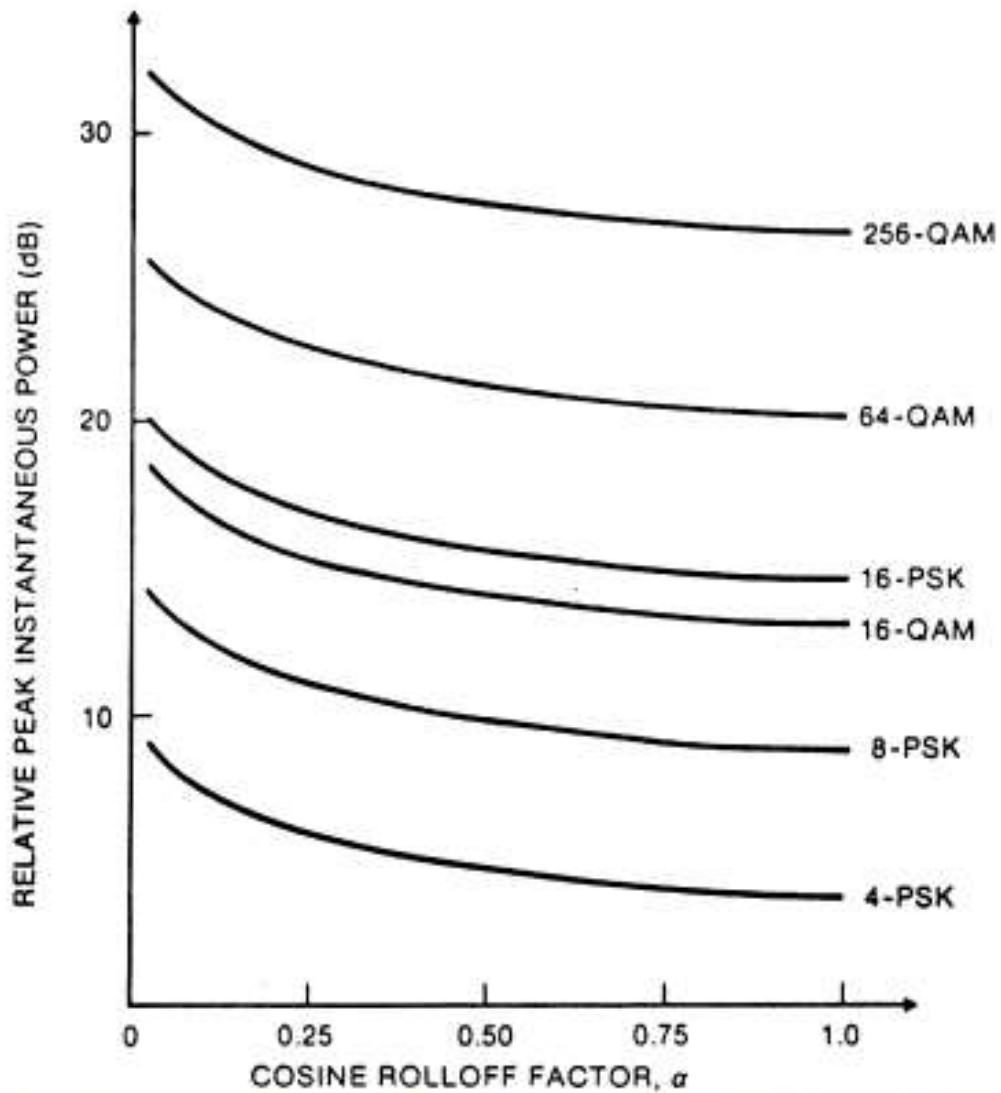


Figure 3.7. Peak instantaneous power (relative to QPSK with square pulses) for M -ary waveforms for several values of M . (From *IEEE Communications Magazine*, Oct. 1986, Figure 8, reprinted with permission; Ref. 4.)

Persyaratan dan sasaran unjuk kerja lintasan radio digital



- BER : perbandingan jml bit error dgn jml bit diterima pd interval waktu
- RBER (Residual BER) : perbandingan bit error tanpa fading, termasuk error system inherent, lingkungan, efek usia dan interferensi jangka-panjang
- ES (Error Second) : perioda 1 detik terjadi 1 atau lebih bit error atau paling tidak terdeteksi 1 error
- SES (Severely Errored Second) : perioda 1 detik terjadi $\text{BER} \geq 1 \times 10^3$ atau paling tidak 1 rusak
- DM (degraded minute) : interval waktu m detik, 60 diantaranya tidak SES tetapi error ratio-nya lebih besar dr yg dispesifikasi

TABLE 3.1 Comparison of Different Modulation Schemes^a

| System | Variants | $W(-E_b/N_0)$ (dB) | S/N (dB) | Nyquist Bandwidth (b_n) |
|---|--|-----------------------|-------------|-----------------------------------|
| Basic Modulation Schemes | | | | |
| FSK | 2-state FSK with discriminator detection | 13.4 | 13.4 | B |
| | 3-state FSK (duo-binary) | 15.9 | 15.9 | B |
| | 4-state FSK | 20.1 | 23.1 | B/2 |
| PSK | 2-state PSK with coherent detection | 10.5 | 10.5 | B |
| | 4-state PSK with coherent detection | 10.5 | 13.5 | B/2 |
| | 8-state PSK with coherent detection | 14.0 | 18.8 | B/3 |
| QAM | 16-state PSK with coherent detection | 18.4 | 24.4 | B/4 |
| | 16-QAM with coherent detection | 17.0 | 20.5 | B/4 |
| | 32-QAM with coherent detection | 18.9 | 23.5 | B/5 |
| QAM | 64-QAM with coherent detection | 22.5 | 26.5 | B/6 |
| | 128-QAM with coherent detection | 24.3 | 29.5 | B/7 |
| | 256-QAM with coherent detection | 27.8 | 32.6 | B/8 |
| QPR ^b | 512-QAM with coherent detection | 28.9 | 35.5 | B/9 |
| | 9-QPR with coherent detection | 13.5 | 16.5 | B/2 |
| | 25-QPR with coherent detection | 16.0 | 20.8 | B/3 |
| | 49-QPR with coherent detection | 17.5 | 23.5 | B/4 |
| Basic Modulation Schemes with Forward Error Correction | | | | |
| QAM with block codes ^c | 16-QAM with coherent detection | 13.9 | 17.6 | $B/4 \times (1+r)$ |
| | 32-QAM with coherent detection | 15.6 | 20.6 | $B/5 \times (1+r)$ |
| | 64-QAM with coherent detection | 19.4 | 23.8 | $B/6 \times (1+r)$ |
| | 128-QAM with coherent detection | 21.1 | 26.7 | $B/7 \times (1+r)$ |
| | 256-QAM with coherent detection | 24.7 | 29.6 | $B/8 \times (1+r)$ |
| | 512-QAM with coherent detection | 25.6 | 32.4 | $B/9 \times (1+r)$ |

^aTheoretical W and S/N values at 10^{-6} BER; calculated values may differ slightly due to different assumptions.

^bQPR = quadrature partial response.

^cAs an example, BCH error correction with a redundancy of 6.7% ($r = 6.7\%$) is used for calculations in this table.

Source: Table 1a, p. 241, ITU-R Rec. F.1101, 1994 (Rev.1).

$$RSL_{BER} = 10 \log kT + 10 \log b_n + NF_{(dB)} + \left(\frac{S}{N}\right)_{BER}$$

RSL_{BER} : aras sinyal terima

K : Konstanta Boltzman = $1,38 \times 10^{-23}$ J/oK

T : Suhu noise oK

b_n : laju symbol

NF : Noise Figure

(SN)_{BER} : perbandingan sinyal thd noise

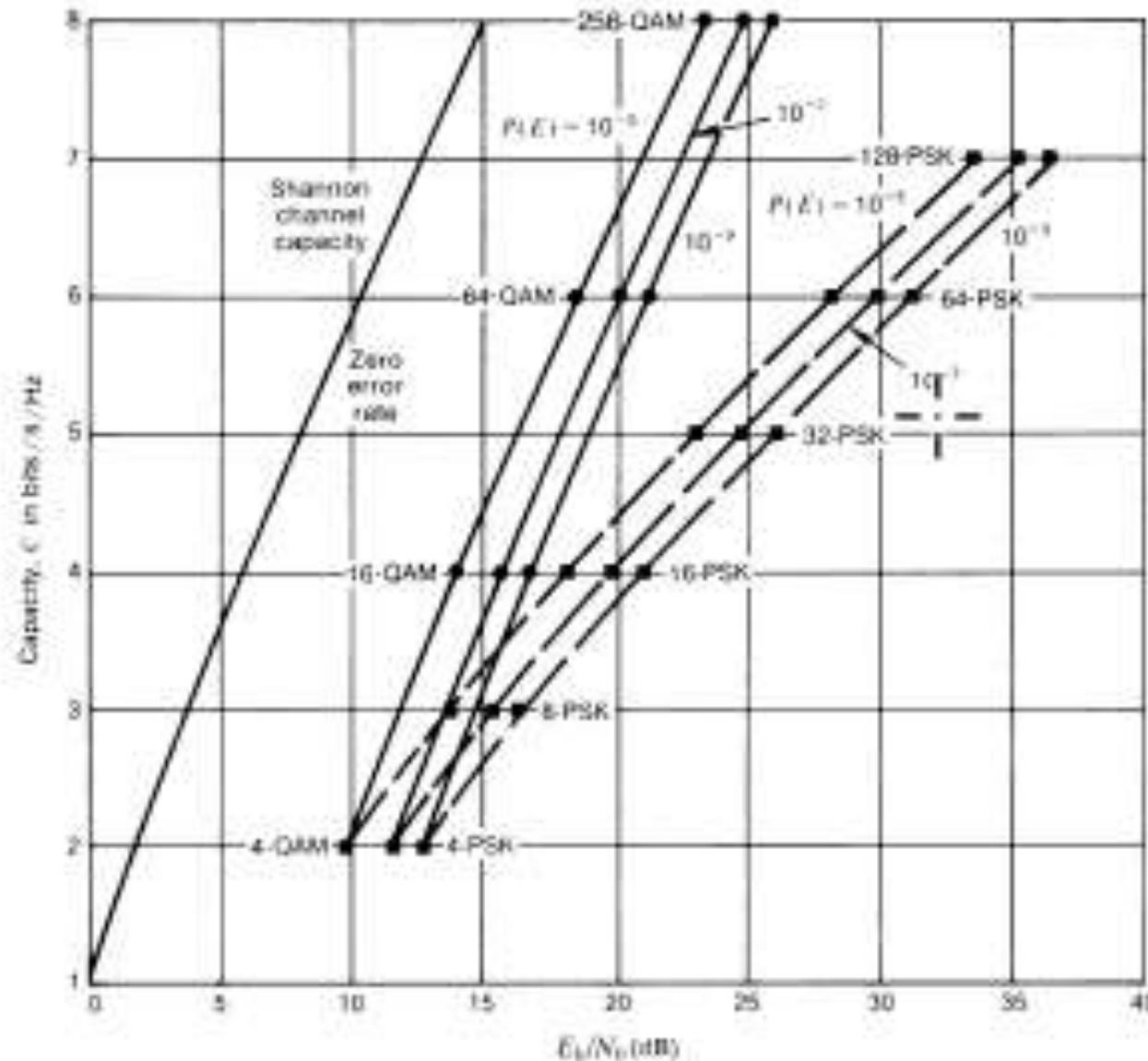


Figure 3.11. Channel capacity comparisons for ideal M -QAM and M -PSK systems (Ref. 8).

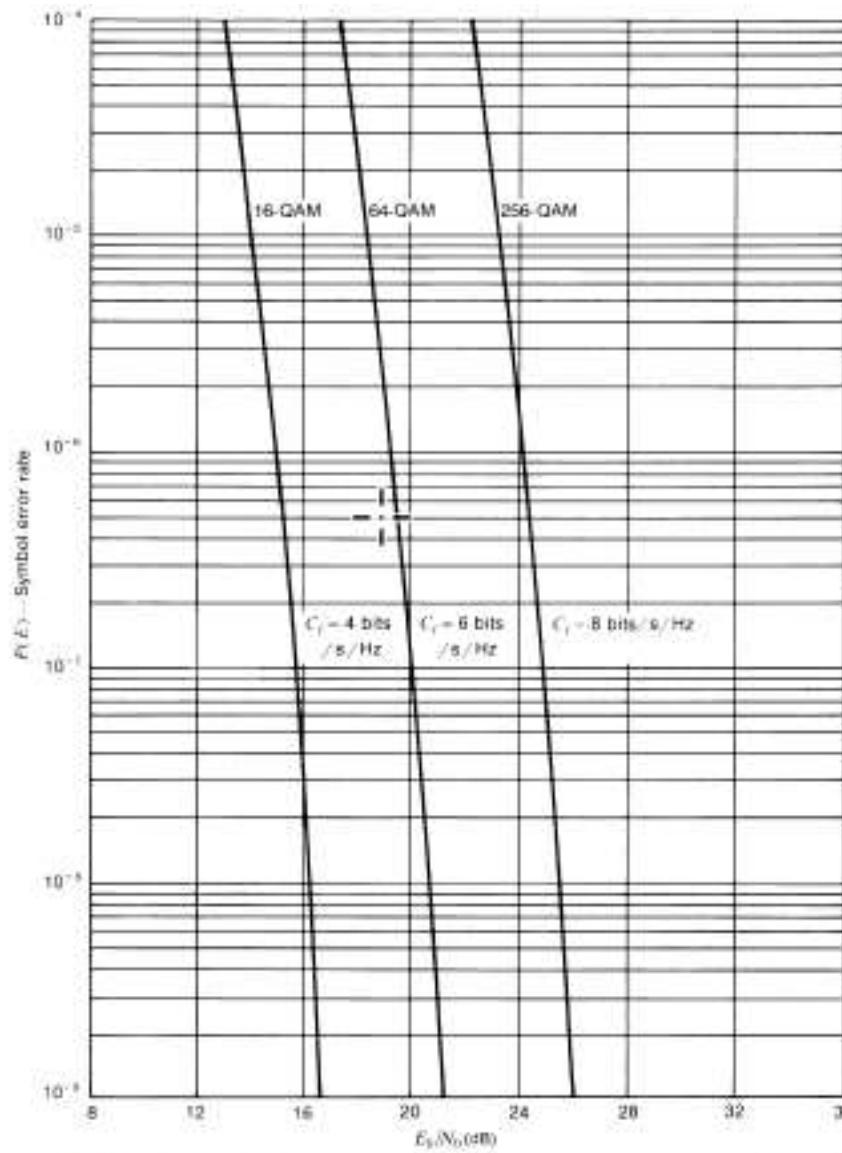


Figure 3.12. Ideal M-QAM performance. (From Ref. 8.) Note: Symbol error rate (SER) can be converted to bit error rate (BER) assuming Gray coded state assignment by:

$$\text{BER} = (1/\log M) \times \text{SER}$$